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# 1-Input/4-Output **Video Distribution Amplifiers**

## **General Description**

The MAX4137/MAX4138 are 1-input/4-output voltagefeedback amplifiers that combine high speed with fast switching for video distribution applications. The MAX4137 is internally set for a closed-loop gain of 2V/V, while the MAX4138 can be externally set for gains of 2V/V or greater.

The MAX4137 achieves a -3dB bandwidth of 185MHz, with 0.1dB gain flatness to 40MHz. The MAX4138's -3dB bandwidth is 140MHz, with 0.1dB gain flatness to 40MHz. Both devices deliver a 1000V/µs slew rate, as well as exceptional full-power bandwidths of 185MHz and 140MHz, respectively.

A 25ns channel switching time enables rapid multiplexing for picture-in-picture applications, yet maintains a high off-isolation of 75dB and all-hostile crosstalk of -50dB (f = 30MHz). The MAX4137/MAX4138's on-board logic selects any combination of the four signal outputs. Each output is capable of swinging ±2V and delivering up to 65mA of current.

For applications that require a 1-input/6-output distribution amplifier, see the MAX4135/MAX4136 data sheet.

## **Selector Guide**

PART	NO. OF OUTPUTS	GAIN (V/V)	-3dB BANDWIDTH (MHz)		
MAX4135	6	Fixed 2	185		
MAX4136	6	≥2	140		
MAX4137	4	Fixed 2	185		
MAX4138	4	≥2	140		

## **Applications**

Video Switching and Distribution High-Resolution RGB CRT Monitors High-Speed Analog Bus Drivers RF Signal Processing Composite Video Preamplifiers

## **Features**

- ♦ Fixed Gain of 2V/V (MAX4137) **External Gain Set (MAX4138)**
- ♦ High Speed:

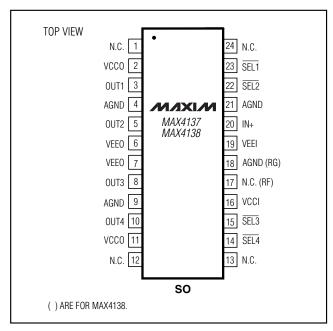
185MHz -3dB Bandwidth (MAX4137) 140MHz -3dB Bandwidth (MAX4138) 1000V/µs Slew Rate

- ♦ High Full-Power Bandwidths (Vout = 2Vp-p): 185MHz (MAX4137) 140MHz (MAX4138)
- ♦ 0.1dB Gain Flatness to 40MHz
- ♦ Low Differential Gain/Phase Error: 0.10%/0.02°
- **♦ High-Impedance Output Disable**

## **Ordering Information**

PART	TEMP. RANGE	PIN-PACKAGE
MAX4137EWG	-40°C to +85°C	24 Wide SO
MAX4138EWG	-40°C to +85°C	24 Wide SO

## Pin Configuration



MIXIM

Maxim Integrated Products 1

## **ABSOLUTE MAXIMUM RATINGS**

Power-Supply Voltage (VCC to VEE)	12V
Voltage on Any Input Pin to GND(Vcc +	0.3V) to (VEE - 0.3V)
Short-Circuit Duration to GND	Continuous
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
Wide SO (derate 19.3mW/°C above +70°C	C)1.54W

Operating Temperature Range	
MAX4137EWG/MAX4138EWG	40°C to +85°C
Storage Temperature Range	65°C to +160°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +5V, V_{EE} = -5V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$ 

PARAMETER	SYMBOL	CC	MIN	TYP	MAX	UNITS	
Input Offset Voltage	Vos	$V_{OUT} = 0V, R_L = 1$		1	10	mV	
Input Offset Voltage Match Between Channels		V <sub>OUT</sub> = 0V, R <sub>L</sub> = 1		1	8	mV	
Input Offset Voltage Drift	TCVOS	V <sub>OUT</sub> = 0V, R <sub>L</sub> = ∞	,		30		μV/°C
Input Bias Current	IB	V <sub>OUT</sub> = 0V, R <sub>L</sub> = ∞	, V <sub>IN</sub> = -V <sub>OS</sub>		4.5	13	μΑ
Common-Mode Input Resistance	RIN(CM)	MAX4138, either in	put		5		ΜΩ
Common-Mode Input Capacitance	CIN(CM)	MAX4138, either input			2		pF
Input Voltage Noise	_	f = 1MHz			7		nV/√Hz
input voitage noise	en	f = 1MHz to 100MHz			88		μV <sub>RMS</sub>
Input Current Noise	in	f = 1MHz			2.4		pA/√Hz
input Guirent Noise	ın	f = 1MHz to $100MHz$			30		nA <sub>RMS</sub>
Input Capacitance	CIN				2		рF
Common-Mode Input Voltage Range	V <sub>CM</sub>				±2.5		V
Common-Mode Rejection Ratio	CMRR	$V_{CM} = \pm 2.5V$			60		dB
Power-Supply Rejection Ratio	PSRR	$V_S = \pm 4.75V$ to $\pm 5.25V$		55	65		dB
Quiescent Supply Current	Isy	V <sub>IN</sub> = 0V	All channels off		30	48	mA
Quiescent Supply Current		VIV = OA	All channels on		47	70	1 IIIA
Output Voltage Swing	Vout	$R_L = 150\Omega$	Positive	2.2	2.6		V
Output Voltage Swing			Negative	-2.0	-2.5		] v
Output Current Drive	lout	$R_L = 30\Omega$		45	65		mA
SEL High Threshold	VIH					2.0	V
SEL Low Threshold	VIL			0.8			V
SEL Input Current	ISEL				1	5	μΑ

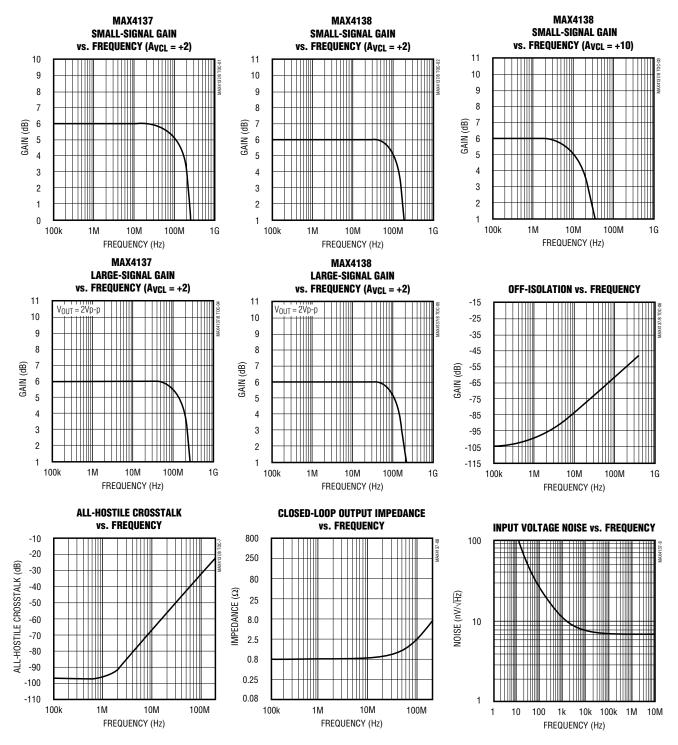
#### **AC ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = +5V, V_{EE} = -5V, A_{VCL} = 2V/V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = 25^{\circ}C.)$ 

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS
-3dB Bandwidth	BW-3dB	V <sub>OUT</sub> ≤ 0.1V <sub>RMS</sub> ,	MAX4137		185		MHz
		$A_{VCL} = 2V/V$	MAX4138		140		IVIMZ
5 H D D I : HI	FPBW	$V_{OUT} = 2V_{p-p}$	MAX4137		185		MHz
Full-Power Bandwidth	FFDW	A <sub>VCL</sub> = 2V/V	MAX4138		140		
0.1dB Bandwidth		AVCL = 2V/V		40		MHz	
Slew Rate	SR	$-2V \le V_{OUT} \le +2V$			1000		V/µs
Cottling Time	+	-1V ≤ V <sub>OUT</sub> ≤ +1V,	to 0.1%		17		ns
Settling Time	ts	$R_L = 150\Omega$ , $A_{VCL} = 2V/V$	to 0.01%		40		
Differential Cain	50	f = 3.58MHz, A <sub>VCL</sub> = 2V/V	MAX4137		0.10		- %
Differential Gain	DG		MAX4138		0.10		
Differential Phase	DP	f = 3.58MHz, A <sub>VCL</sub> = 2V/V	MAX4137		0.02		degrees
Differential Priase			MAX4138		0.02		
All-Hostile Crosstalk		V <sub>IN</sub> = 1Vp-p, f = 30MHz			-50		dB
Off Isolation		$V_{IN} = 1Vp-p, f = 30MHz$			75		dB
Channel Switching Off Time	toff				25		ns
Channel Switching On Time	ton				25		ns
Digital Switching Feedthrough		V <sub>IN</sub> = 0V <sub>DC</sub>			±1		mV
Spurious-Free Dynamic Range	SFDR	$f_C = 5MHz$ , $A_{VCL} = 2V/V$ , $V_{OUT} = 2Vp-p$ , $R_L = 100\Omega$		-72			dBc
Output On-Resistance	Rout	f = DC, A <sub>VCL</sub> = 2V/V		1			Ω
Output Off-Resistance	Rout	$f = DC$ , $A_{VCL} = 2V/V$		$f = DC$ , $A_{VCL} = 2V/V$ 200			kΩ
Output On-Capacitance	Cout(on)	2			pF		
Output Off-Capacitance	Cout(off)				3.5		pF

## **Typical Operating Characteristics**

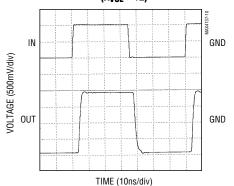
(V<sub>CC</sub> = +5V, V<sub>EE</sub> = -5V, R<sub>L</sub> = 150 $\Omega$ , T<sub>A</sub> = +25°C, unless otherwise noted.)



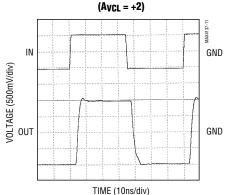
## Typical Operating Characteristics (continued)

(V<sub>CC</sub> = +5V, V<sub>EE</sub> = -5V, R<sub>L</sub> = 150 $\Omega$ , T<sub>A</sub> = +25°C, unless otherwise noted.)

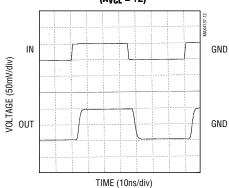
#### MAX4137 LARGE-SIGNAL PULSE RESPONSE (Avcl = +2)



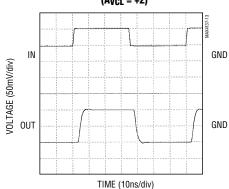
# MAX4138 LARGE-SIGNAL PULSE RESPONSE



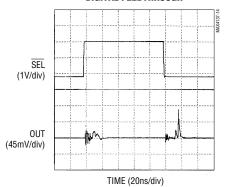
#### MAX4137 SMALL-SIGNAL PULSE RESPONSE (Avcl = +2)



#### MAX4138 SMALL-SIGNAL PULSE RESPONSE (Avgl = +2)

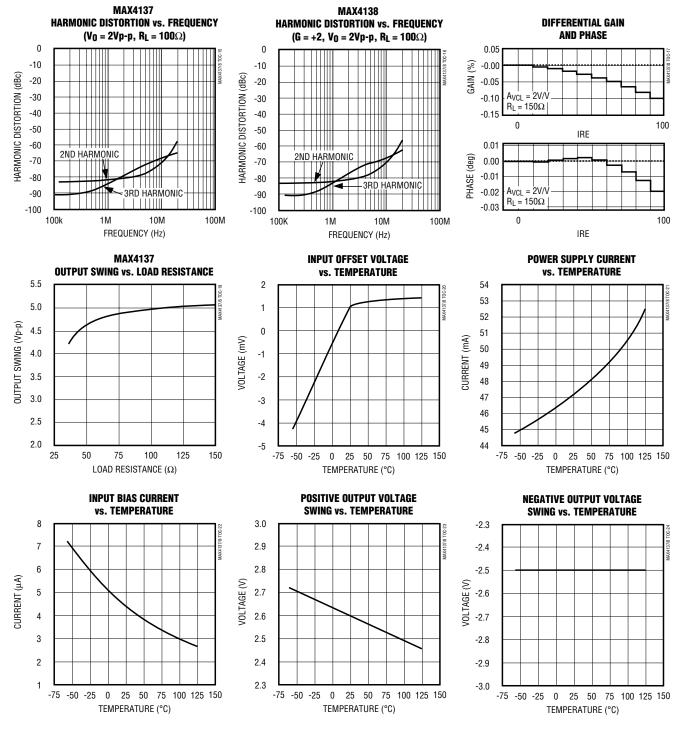


#### DIGITAL FEEDTHROUGH



## Typical Operating Characteristics (continued)

 $(V_{CC} = +5V, V_{EE} = -5V, R_L = 150\Omega, T_A = +25^{\circ}C, unless otherwise noted.)$ 



## Pin Description

PIN		NANT	FUNCTION			
MAX4137	MAX4138	NAME	FUNCTION			
1, 12, 13, 17, 24	1, 12, 13, 24	N.C.	No Connect. Not internally connected.			
2, 11	2, 11	VCCO	Positive Supply for Output Amplifiers. Connect to +5V.			
3	3	OUT1	Output 1			
4, 9, 18, 21	4, 9, 21	AGND	Analog Ground			
5	5	OUT2	Output 2			
6, 7	6, 7	VEEO	Negative Supply for Output Amplifiers. Connect to -5V.			
8	8	OUT3	Output 3			
10	10	OUT4	Output 4			
14	14	SEL4	When low, enables output channel OUT4. When high, disables output channel OUT4.			
15	15	SEL3	When low, enables output channel OUT3. When high, disables output channel OUT3.			
16	16	VCCI	Positive Supply for Input Amplifier. Connect to +5V.			
_	17	RF	Output of Input Amplifier			
_	18	RG	Inverting Input			
19	19	VEEI	Negative Supply for Input Amplifier. Connect to -5V.			
20	20	IN+	Noninverting Input			
22	22	SEL2	When low, enables output channel OUT2. When high, disables output channel OUT2.			
23	23	SEL1	When low, enables output channel OUT1. When high, disables output channel OUT1.			

## **Detailed Description**

The MAX4137/MAX4138 are 1-input/4-output video distribution amplifiers. The MAX4137 is configured for a fixed gain of +2, while the MAX4138 features external gain control (feedback) for closed-loop gains of 2V/V or greater.

Each output provides sufficient current to drive five  $150\Omega$  loads. However, distortion will increase when driving multiple loads. The TTL/CMOS-compatible digital control ( $\overline{SEL}$ ) enables or disables each output amplifier. When the  $\overline{SEL}$  control input is low, the amplifier is enabled; when it is high, the amplifier is disabled and presents a high-impedance output. The enable/disable or

disable/enable time is under 25ns, which is useful in multiplexing, pixel switching, or picture-in-picture applications.

Each device has an input amplifier, which buffers the input from any switching glitches that may be taking place at the output stage, and provides a high-impedance, low-capacitance input. The separate input buffer allows a true high output impedance when an amplifier is disabled.

The outputs are protected against short circuits to ground. However, power-dissipation limits preclude shorting all output channels to ground. See the *Power-Dissipation Considerations* section for details.

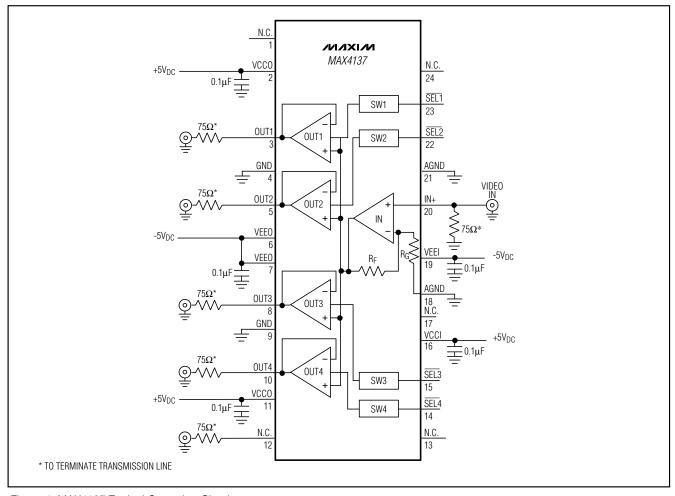


Figure 1. MAX4137 Typical Operating Circuit

## **Applications Information**

# Grounding, Bypassing, and PC Board Layout

To obtain the MAX4137/MAX4138s' full 185MHz bandwidth, Microstrip and Stripline techniques are recommended in most cases. To ensure the PC board does not degrade the amplifier's performance, design the board for a frequency greater than 1GHz. Even with very short traces, use these techniques at critical points, such as inputs and outputs. Whether you use a constant-impedance board or not, observe the following guidelines when designing the board:

- Do not use wire-wrap boards. They are too inductive.
- Do not use IC sockets. They increase parasitic capacitance and inductance.
- In general, surface-mount components have shorter leads and lower parasitic reactance, giving better high-frequency performance than through-hole components.
- The PC board should have at least two layers, with one side a signal layer and the other a ground plane.
- Keep signal lines as short and straight as possible.
   Do not make 90° turns; round all corners.
- The ground plane should be as free from voids as possible.

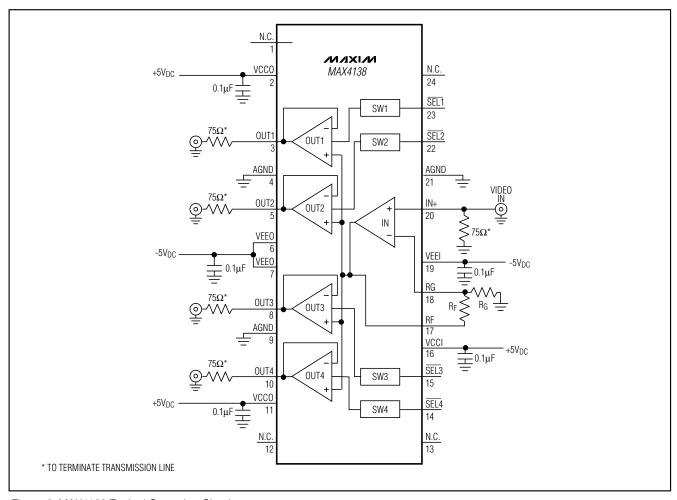


Figure 2. MAX4138 Typical Operating Circuit

#### **Driving Capacitive Loads**

The MAX4137/MAX4138 provide maximum AC performance with no output load capacitance. This is the case when they are driving a correctly terminated transmission line (i.e., a back-terminated 75 $\Omega$  cable). However, the MAX4137/MAX4138 are capable of driving capacitive loads up to 10pF without oscillations, but with reduced AC performance.

Driving large capacitive loads increases the chance of oscillations in most amplifier circuits. This is especially true for circuits with high loop gain, such as voltage followers. The amplifier's output resistance and the load capacitor combine to add a pole and excess phase to

the loop response. If the frequency of this pole is low enough and phase margin is degraded sufficiently, oscillations may occur.

A second problem when driving capacitive loads results from the amplifier's output impedance, which looks inductive at high frequencies. This inductance forms an L-C resonant circuit with the capacitive load, which causes peaking in the frequency response and degrades the amplifier's gain margin.

The MAX4137/MAX4138 drive capacitive loads up to 10pF without oscillation. However, some peaking (in the frequency domain) or ringing (in the time domain) may occur (Figure 3).

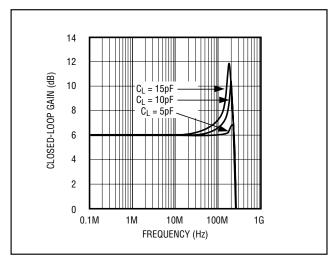


Figure 3. Effect of  $C_{LOAD}$  on Frequency Response (without RISO)

To drive larger-capacitance loads or to reduce ringing, add an isolation resistor between the amplifier's output and the load, as shown in Figure 4.

The value of R<sub>ISO</sub> depends on the circuit's gain and the capacitive load. Figure 5 shows the optional isolation resistor (R<sub>ISO</sub>) vs. capacitive load ( $C_L$ ). At the higher capacitor values, the bandwidth is dominated by the RC network, formed by R<sub>ISO</sub> and  $C_L$ .

#### **Power-Dissipation Considerations**

The MAX4137/MAX4138 can drive up to four outputs simultaneously. Quiescent power dissipation is typically 520mW and 650mW maximum, respectively, with all channels enabled. The maximum package power dissipation is rated at 1540mW.

In a typical application, four outputs drive a standard video signal into a  $150\Omega$  load. The amount of power added to the quiescent dissipation is minimal and no special precautions are necessary.

However, each output driving the maximum 65mA into  $30\Omega$  will cause a power-dissipation increase of approximately 200mW. Therefore, you should not allow more than three outputs to deliver that load simultaneously. Similarly, one output shorted to ground will cause a power-dissipation increase of 650mW. Only one output can be shorted to ground without violating the package power rating.

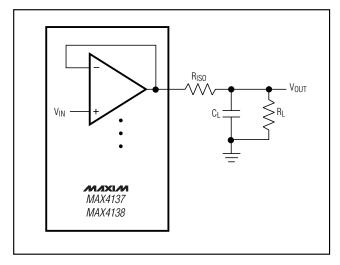


Figure 4. Capacitive-Load Driving Circuit

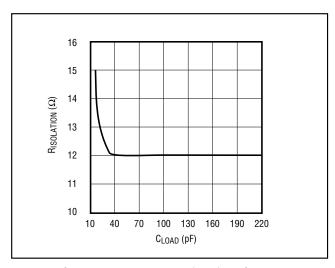


Figure 5. Optimal Isolation Resistor (RISO) vs. CLOAD

In conclusion, during normal operation in a matchedload environment, the total power dissipation is well within the package's dissipation rating. The maximum power dissipation is violated only if multiple channels are driving the maximum current into minimum loads at the same time.

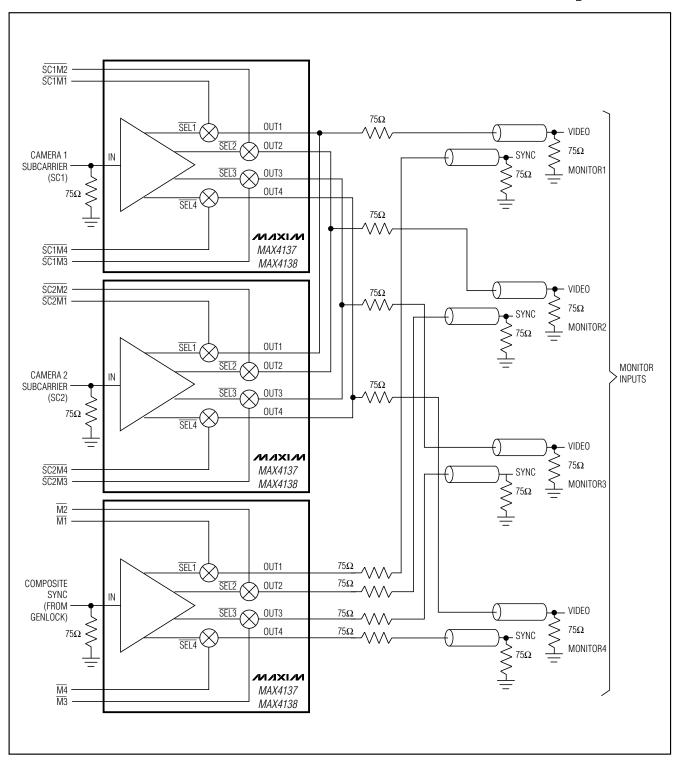


Figure 6. Two Cameras to Four Monitors Distribution Amplifier

**Chip Information** 

TRANSISTOR COUNT: 625 SUBSTRATE CONNECTED TO VEE

## Package Information

**MILLIMETERS** 

MIN

2.35

0.10

0.35

0.23

7.40

10.00

0.40

MAX

2.65

0.30

0.49

0.32

7.60

10.65

1.27

10.50

11.75

13.00

15.60

18.10 21-0042A

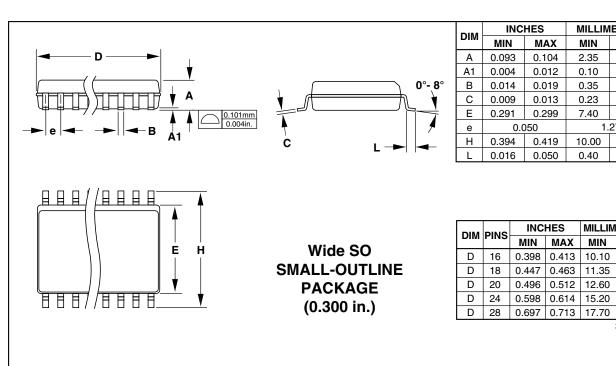
**MILLIMETERS** 

MIN

10.10

11.35

12.60



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