

(12.91)

$$X_{L1} = \omega L_1 = \frac{R_s(Q_L^2 + 1)}{Q_L - \sqrt{\frac{R_s(Q_L^2 + 1)}{R_i}} - 1}. \quad (12.98)$$

This circuit can match resistances

$$\frac{R_i}{Q_L^2 + 1} < R_s < R_i. \quad (12.99)$$

Suboptimum operation takes place for

$$R_{s(sub)} > R_s \quad (12.100)$$

and therefore for

$$R_{i(sub)} < R_i. \quad (12.101)$$

(12.92)

12.7 DESIGN EXAMPLE

EXAMPLE 12.1

(12.93)

Design the Class E ZVS inverter of Fig. 12.1(a) to satisfy the following specifications: $V_I = 100$ V, $P_{Rimax} = 80$ W, and $f = 1.2$ MHz. Assume $D = 0.5$.

Solution: It is sufficient to design the inverter for the full power. Using (12.48), the full-load resistance is

(12.94)

$$R_i = \frac{8}{\pi^2 + 4} \frac{V_I^2}{P_{Ri}} = 0.5768 \times \frac{100^2}{80} = 72.1 \, \Omega. \quad (12.102)$$

(12.95)

From (12.41), the DC resistance of the inverter is

(12.96)

$$R_{DC} = \frac{\pi^2 + 4}{8} R_i = 1.7337 \times 72.1 = 125 \, \Omega. \quad (12.103)$$

The amplitude of the output voltage is computed from (12.46)

$$V_{Rim} = \frac{4}{\sqrt{\pi^2 + 4}} V_I = 1.074 \times 100 = 107.4 \, \text{V}. \quad (12.104)$$

The maximum voltage across the switch and the shunt capacitor can be calculated from (12.43) as

(12.97)

$$V_{SM} = 3.562 V_I = 3.562 \times 100 = 356.2 \, \text{V}. \quad (12.105)$$